

Frequency Shift Acceleration Control for Anti-islanding of a Distributed Generator

2008. 9. 9

Seul-Ki Kim

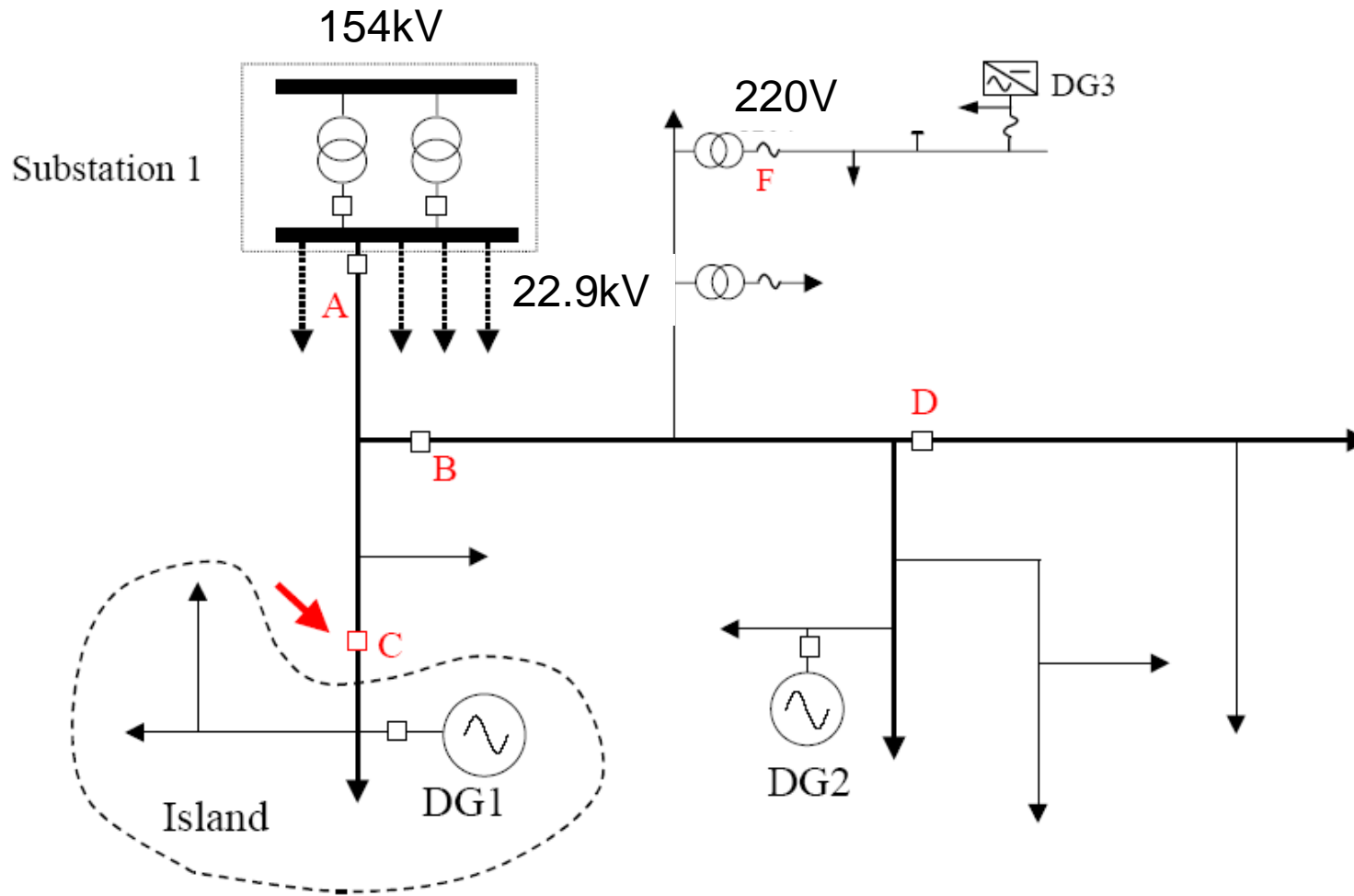
Overview

1. Introduction
2. Frequency Shift Acceleration Control
3. Design of Acceleration Gain
4. Simulation Results
5. Experimental Results
6. Conclusion

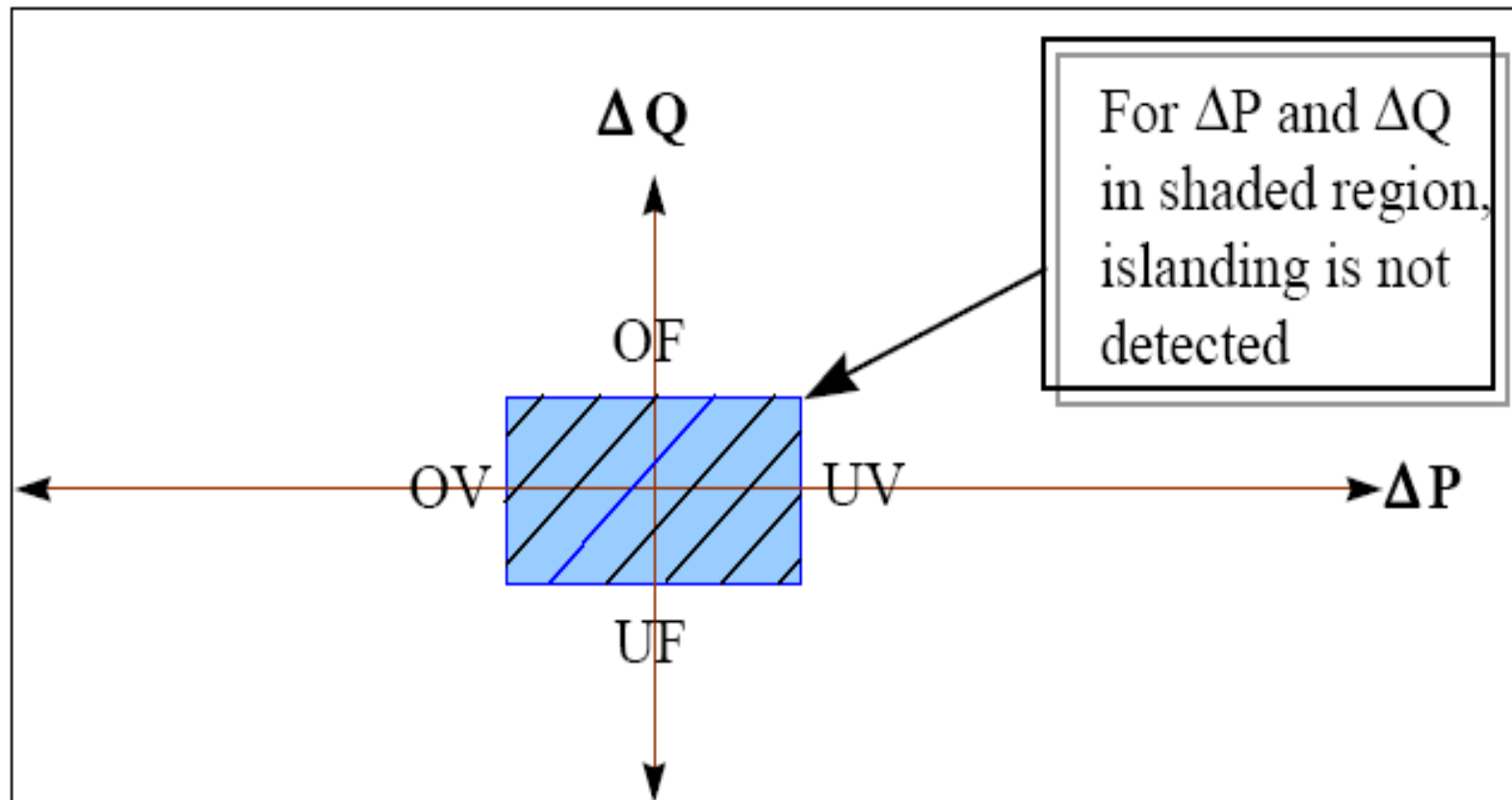
Introduction

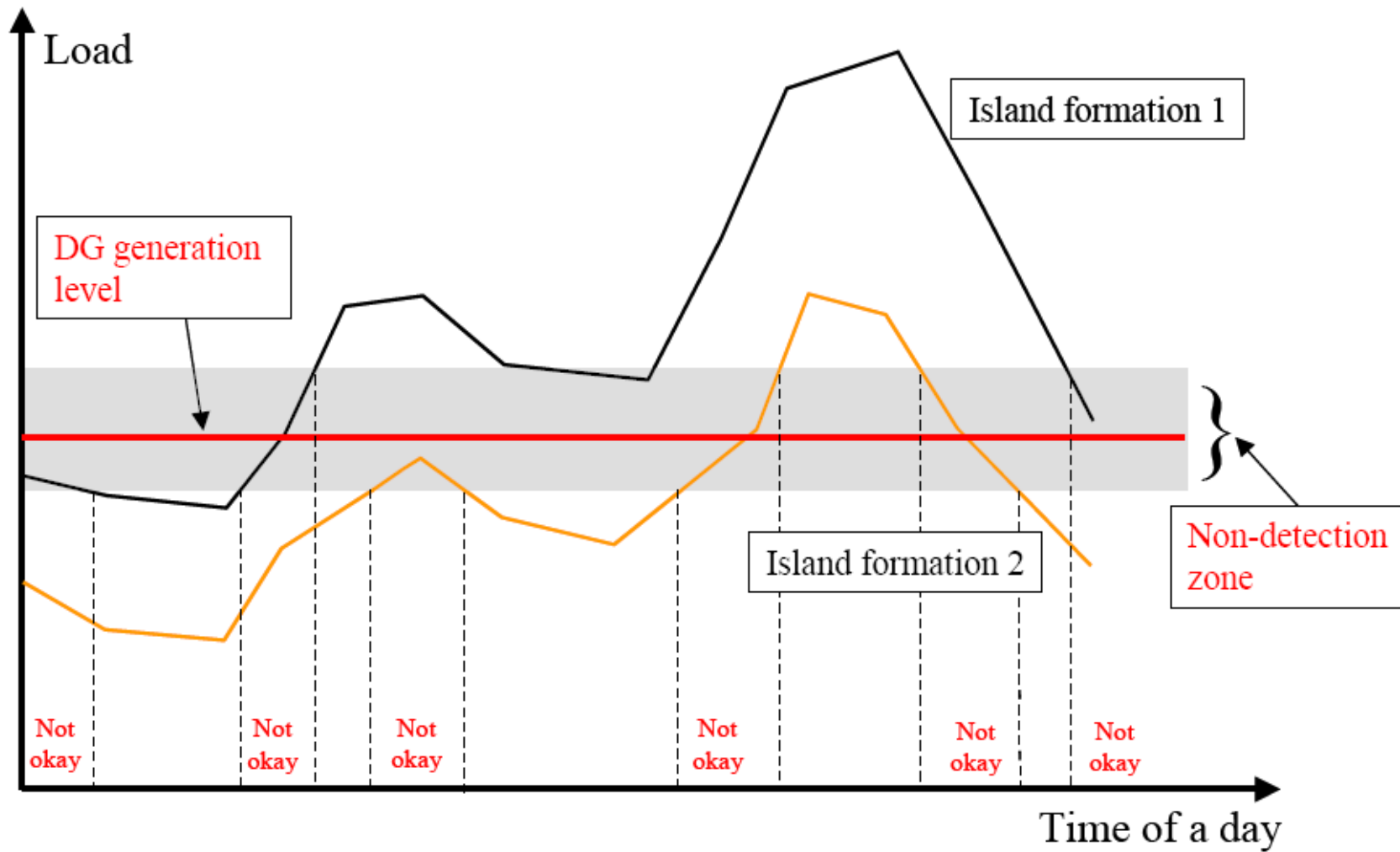
- Islanding
- Anti-islanding Methods
 - Passive
 - Active
- Proposed Algorithm
 - DQ control
 - Small signal analysis
 - Simulation & Experiment

Islanding



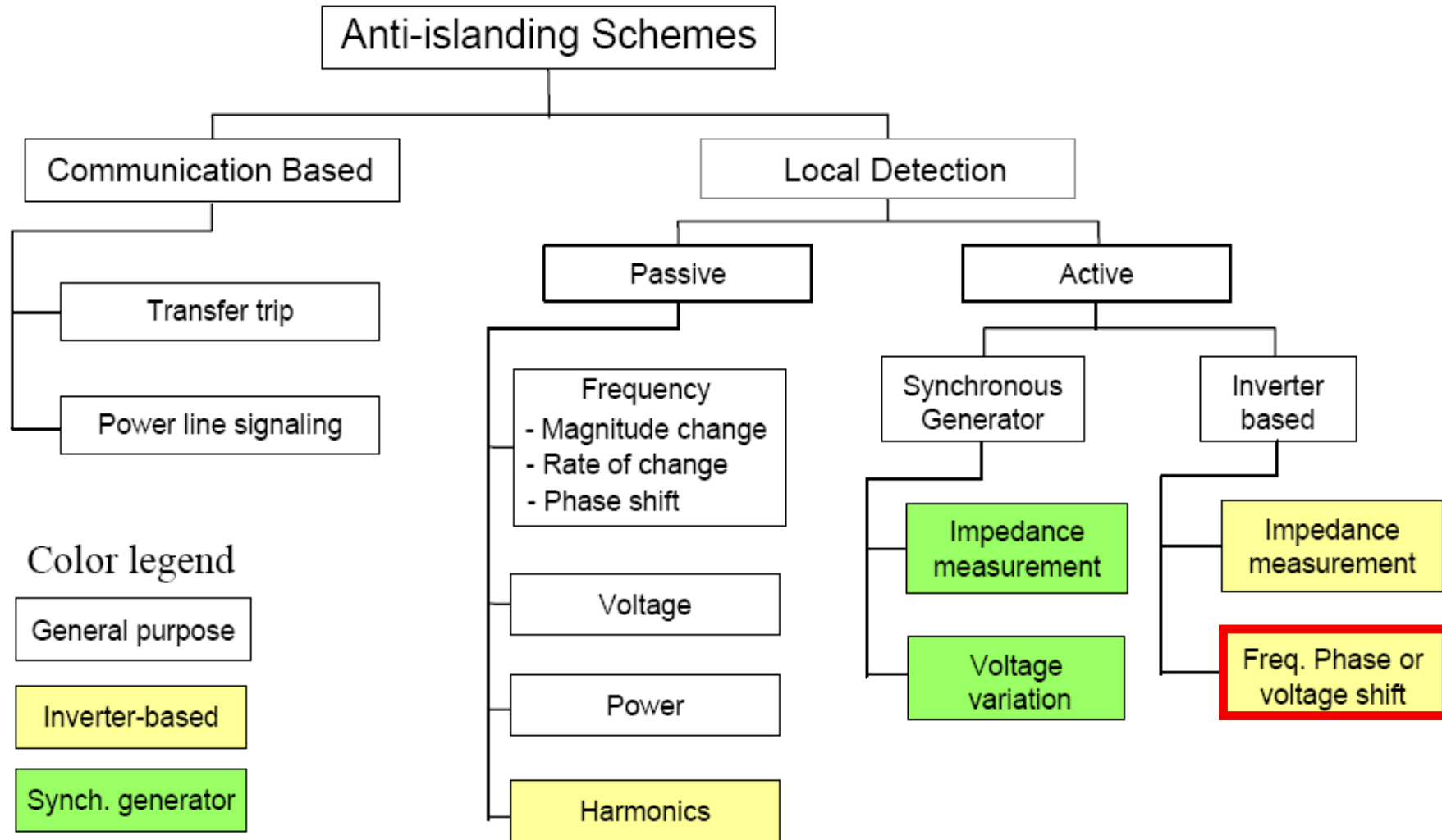
Non-Detection Zone (NDZ)





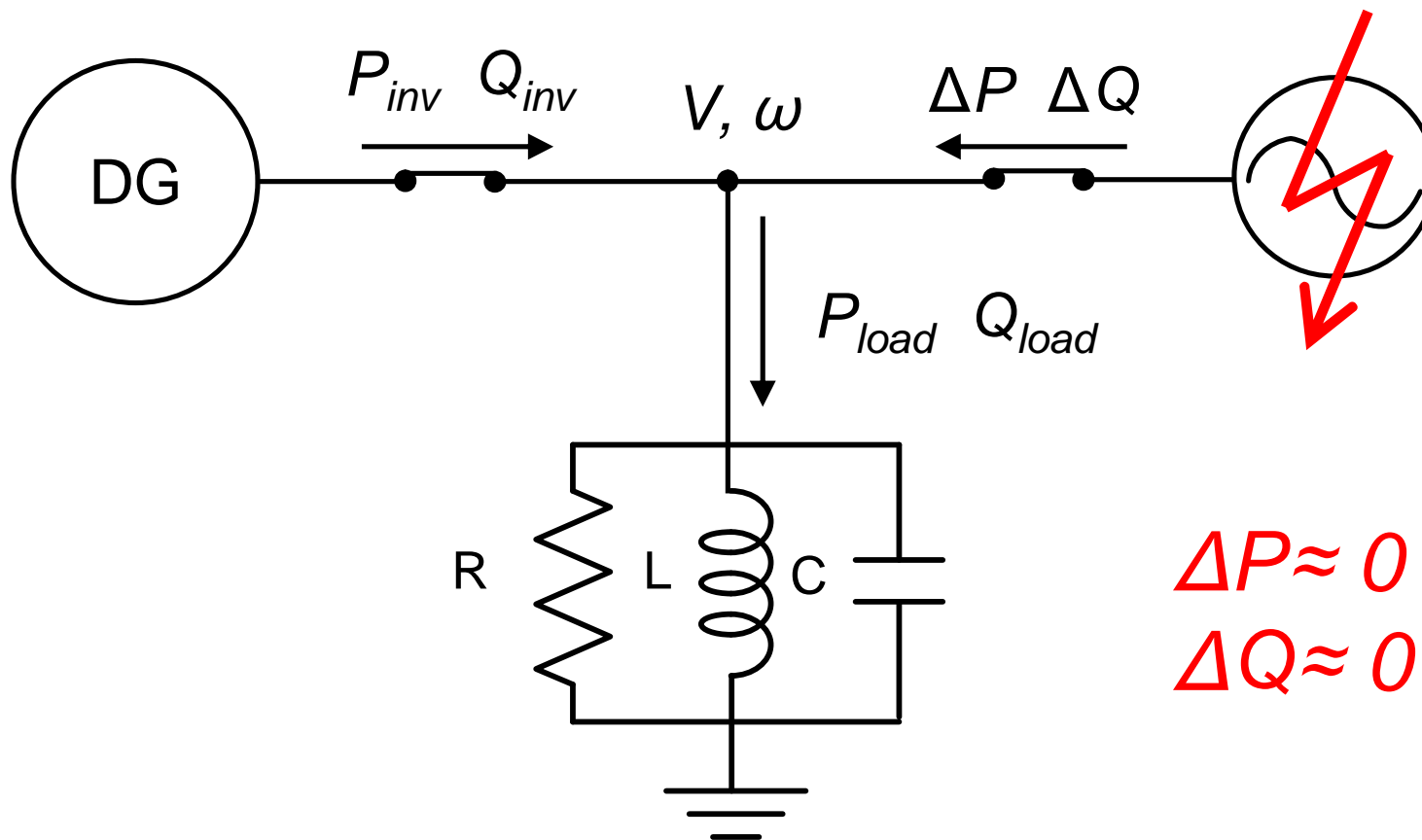
NDZ Impacts Islanding Detection

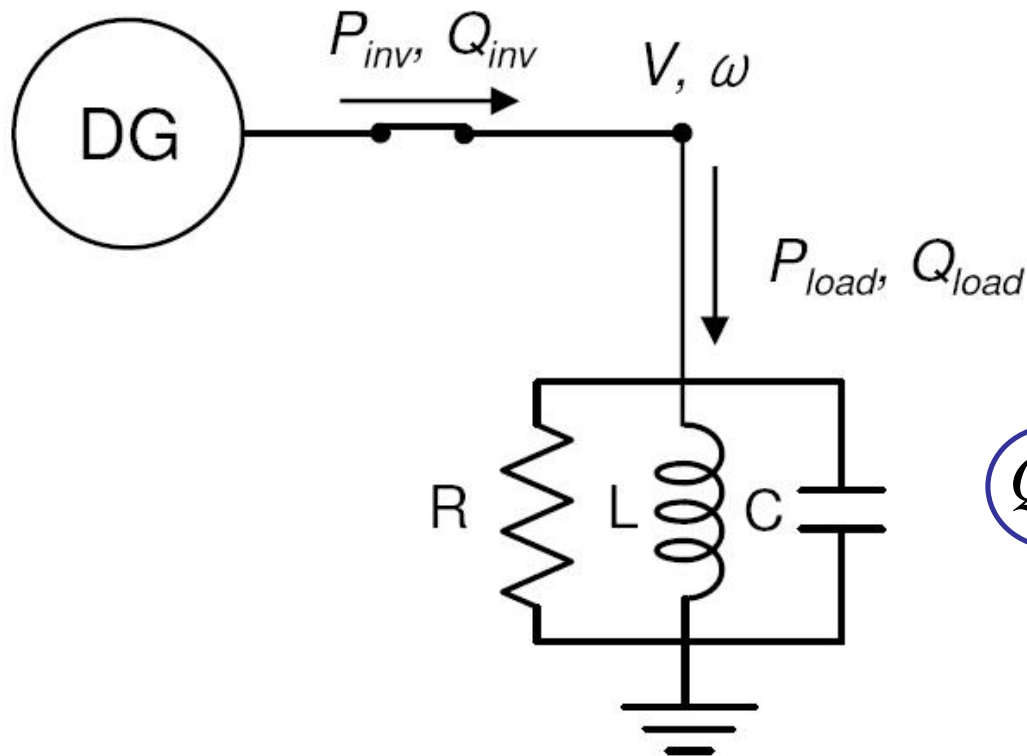
Anti-islanding Methods



FSAC

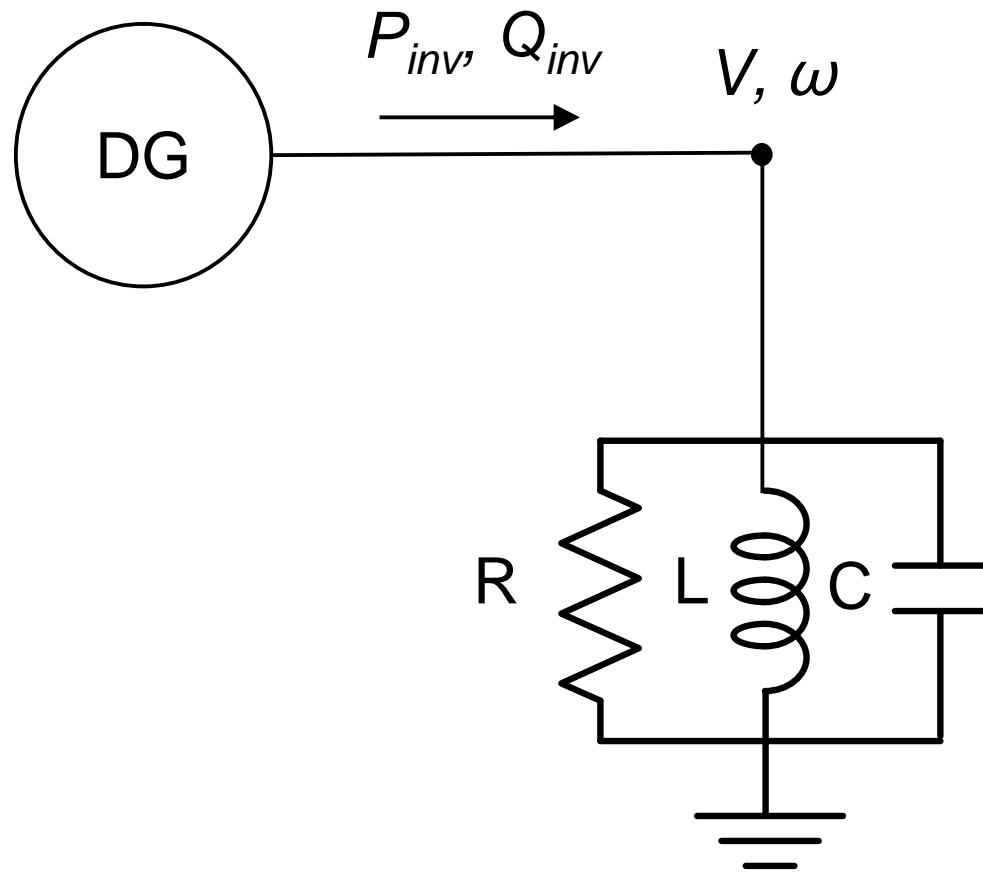
- Islanding Condition





$$P_{inv} = P_{load} = \frac{V^2}{R}$$

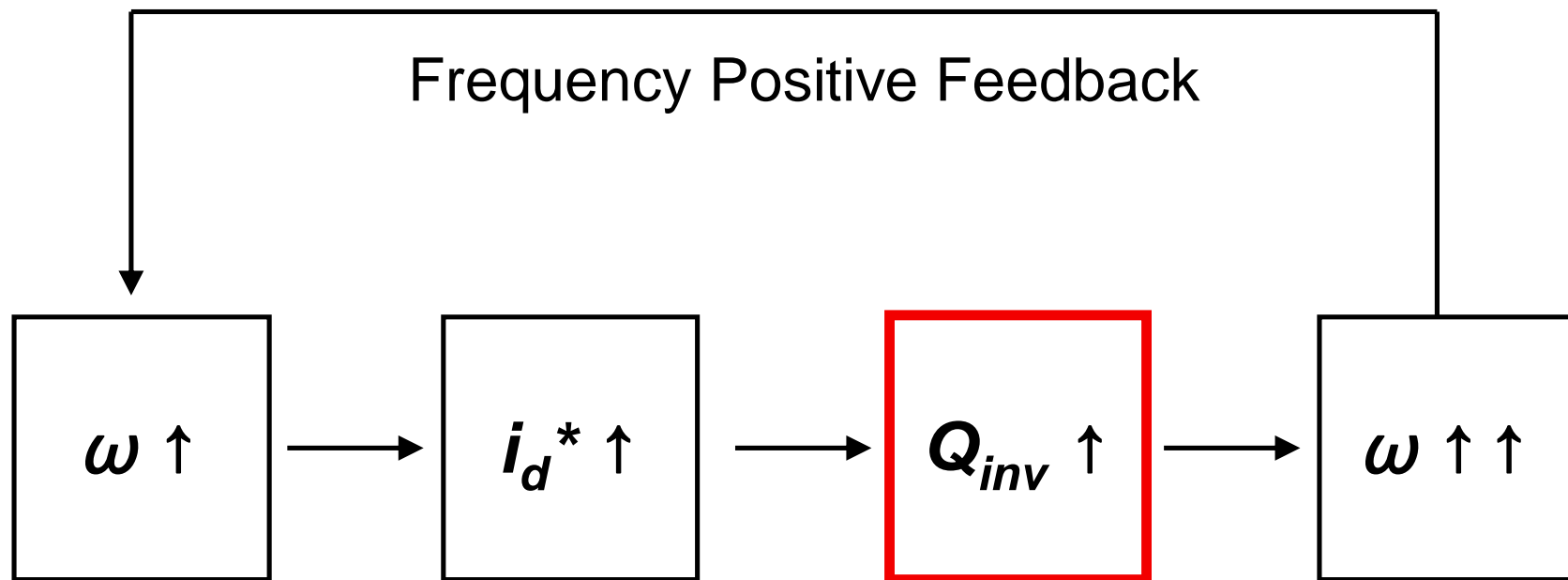
$$Q_{inv} = Q_{load} = V^2 \left(\frac{1}{\omega L} - \omega C \right)$$



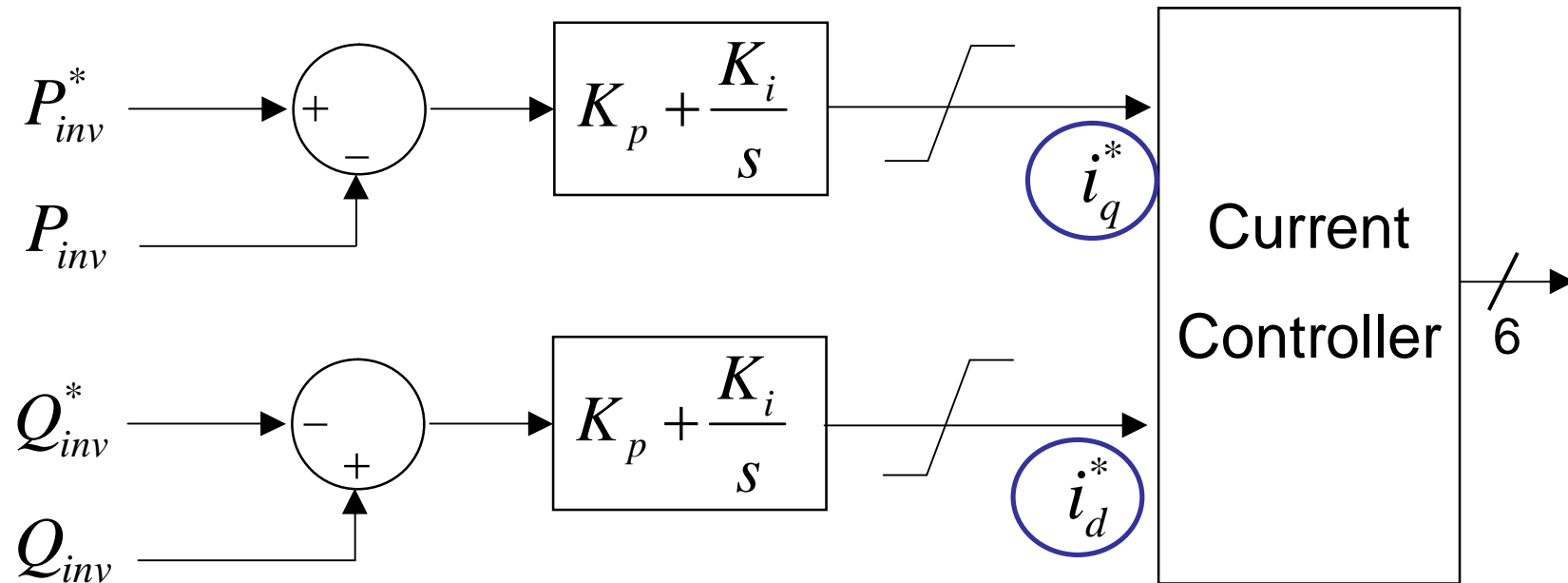
$$P_{inv} \uparrow \rightarrow V \uparrow$$

$$Q_{inv} \uparrow \rightarrow \omega \downarrow$$

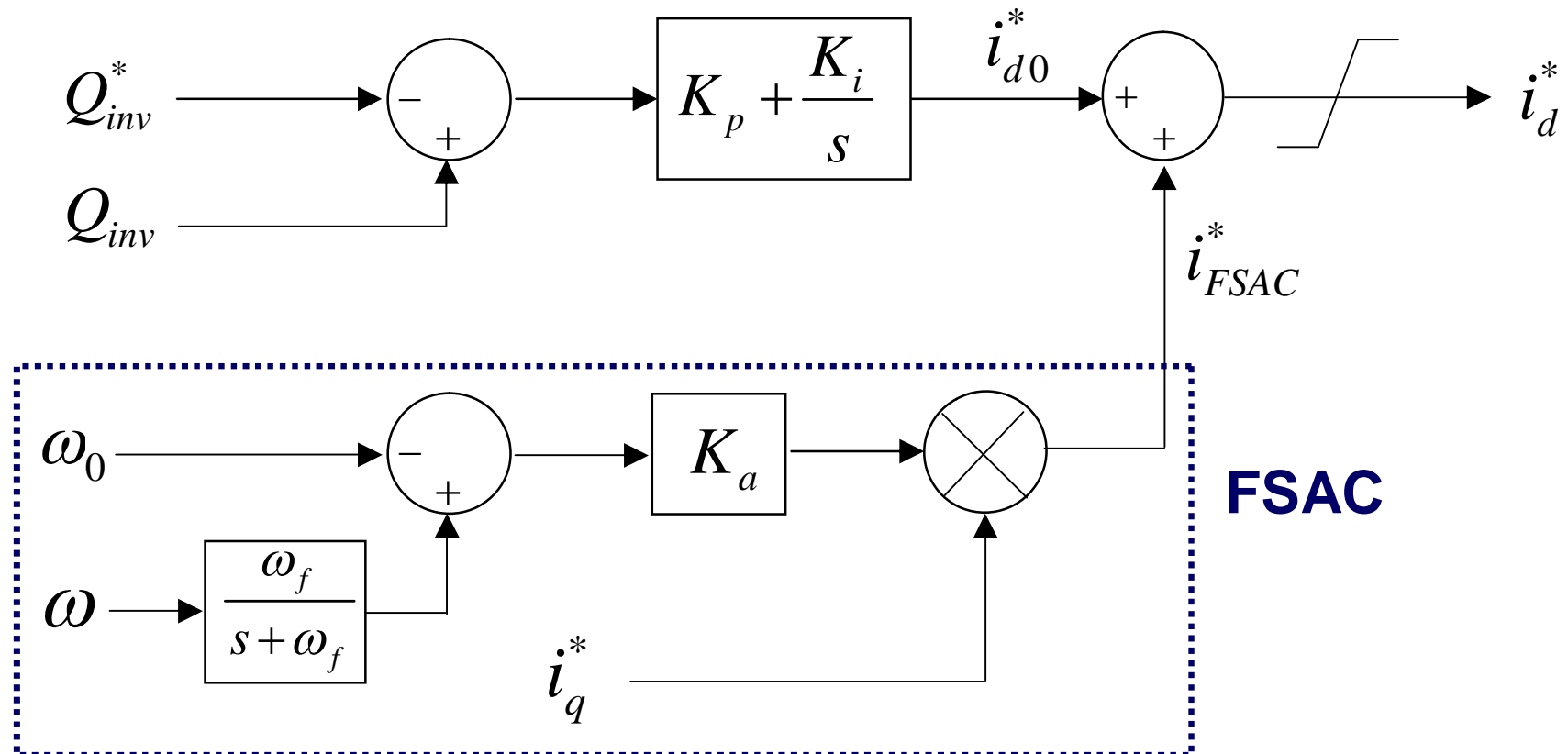
- Key Idea



- Power controller of DG inverters



■ Q controller with FSAC



$$K_{pf} = i_q^* \cdot K_a$$

K_{pf} : Positive feedback gain

K_a : Frequency shift acceleration gain

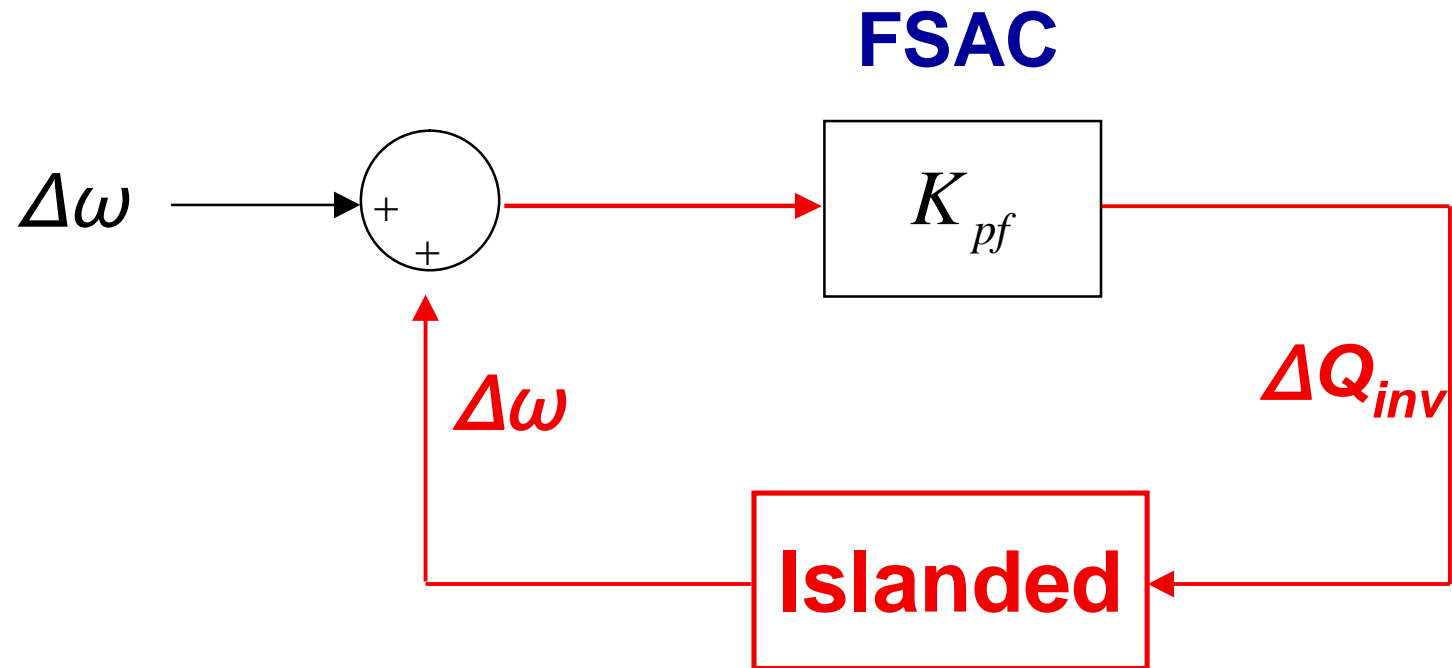
Design of Acceleration Gain

- When islanded, **large** enough to destabilize system
Small Signal Analysis

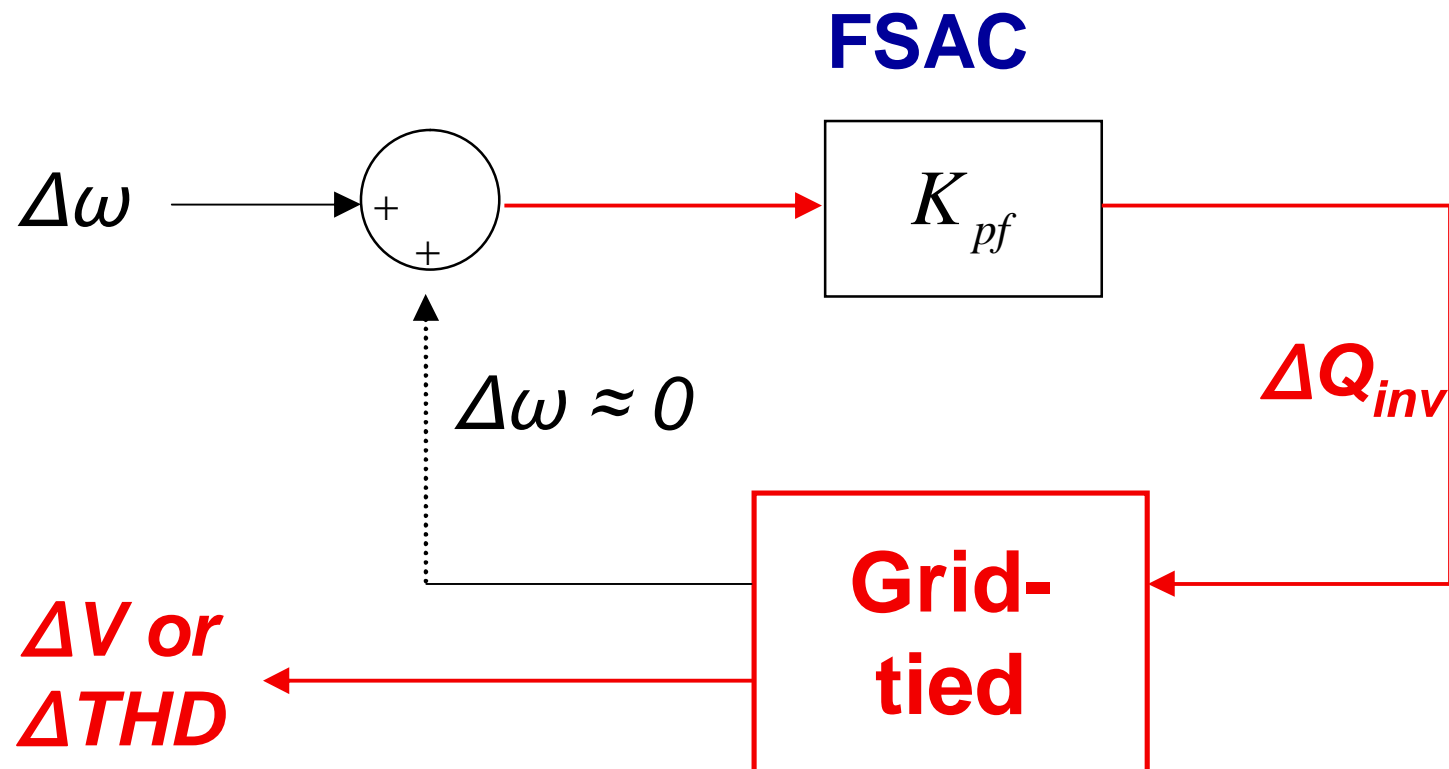
- When grid-tied, **small** enough to keep ΔQ beyond the limit

Frequency Step Response

- $K_{pf} > \text{Lower Limit}$



- $K_{pf} < \text{Upper Limit}$



- Lower Limit by Small Signal Analysis

$$(Q_{inv} - Q^*) \left(K_P + \frac{K_i}{s} \right) + \left(\omega \frac{\omega_f}{s + \omega_f} - \omega_0 \right) K_{pf} = i_d^*$$

$$\left(K_P + \frac{K_i}{s} \right) \Delta Q_{inv} + K_{pf} \frac{\omega_f}{s + \omega_f} \Delta \omega = \Delta i_d$$

$$s^2 + \left[\frac{e_q \left\{ 2 + 3e_q \left(K_P + \frac{K_i}{\omega_f} \right) \right\} \left(\frac{Q_f}{\omega_0 R} \right) - K_{pf}}{2e_q \left(\frac{Q_f}{\omega_0 R} \right) \left(1 + \frac{3}{2} e_q K_P \right)} \right] \omega_f s + \frac{3}{2} e_q \left(\frac{K_i \omega_f}{1 + \frac{3}{2} e_q K_P} \right) = 0$$

For the islanded system to be unstable

$$K_{pf} > \left\{ 2 + 3e_q \left(K_p + \frac{K_i}{\omega_f} \right) \right\} \cdot \left(\frac{Q_f}{\omega_0} \right) \cdot \left(\frac{e_q}{R} \right) \quad i_q$$

$$K_a > \left\{ 2 + 3\sqrt{2}V_n \left(K_p + \frac{K_i}{\omega_f} \right) \right\} \cdot \left(\frac{Q_f}{\omega_0} \right)$$

V_n : Inverter terminal voltage,

ω_f : Measuring frequency

Q_f : Quality factor,

ω_0 : Nominal frequency

**FSAC eliminates
real power dependency of control gain !!**

- Upper Limit by Freq. Step Response

$$\Delta i_d(s) = \left(K_p + \frac{K_i}{s} \right) \Delta Q_{inv}(s) + K_{pf} \Delta \omega(s)$$

$$\left| \Delta Q_{inv}(s) \right| = \frac{K_{pf}}{K_p + 2/(3e_q) + K_i/s} \left| \frac{\Delta \omega}{s} \right|$$

$$\Delta Q_{inv}(t) = \frac{K_{pf}}{K_p + 2/(3e_q)} |\Delta \omega| \exp[st]$$

Maximum Q disturbance due to frequency step change

$$\Delta Q_{\max} > \frac{K_{pf}}{K_p + 2/(3e_q)} |\Delta\omega_{\max}|$$

$$\eta_{\text{preset}} = \frac{\Delta Q_{\max}}{P_{\text{inv}}}$$

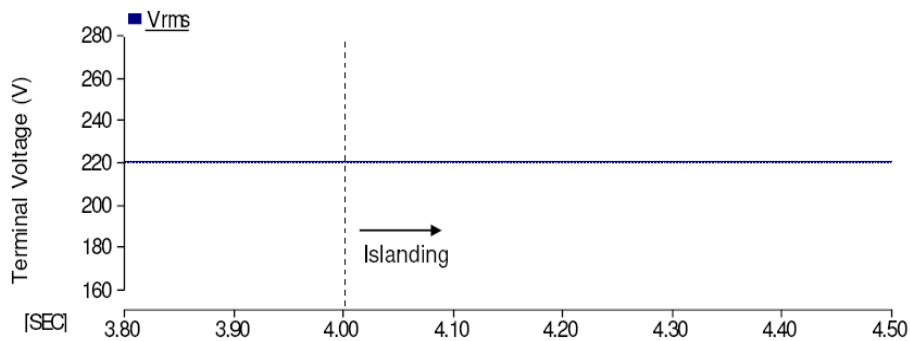
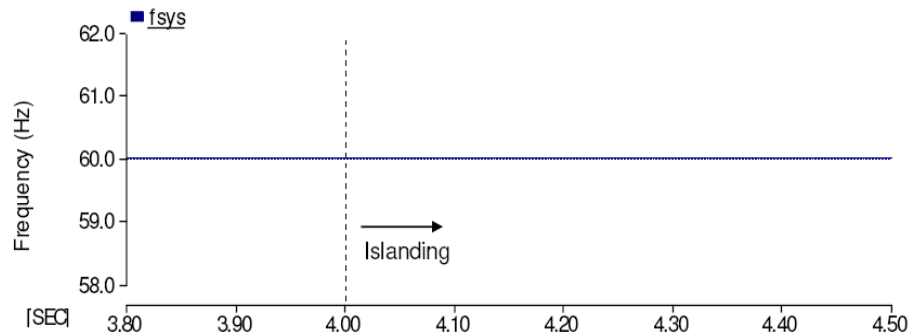
$$K_{pf} < \left(1 + \frac{3}{2} e_q K_p \right) \frac{\eta_{\text{preset}}}{|\Delta\omega_{\max}|} \cdot \dot{i}_q^*$$

$$K_a < \left(1 + \frac{3\sqrt{2}}{2} V_n K_p \right) \frac{1}{|\Delta\omega_{\max}|} \cdot \eta_{\text{preset}}$$

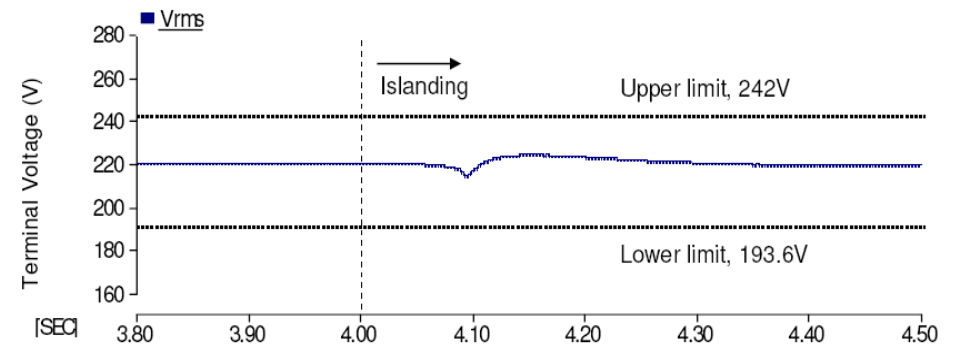
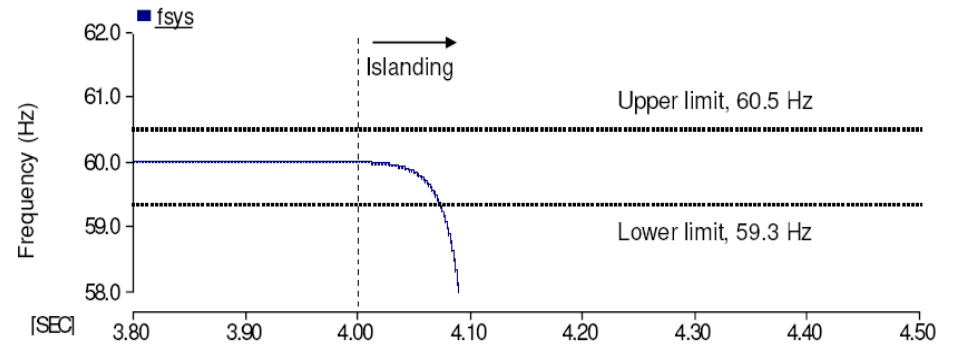
Simulation Results

- Simulation conditions
 - $P_{\text{inv}} = P_{\text{load}} = 20\text{kW}$, $Q_{\text{inv}} = Q_{\text{load}} = 0\text{kVar}$
 - Detection condition (IEEE P1547)
 - Voltage : 110% > or < 88%
 - Frequency : 60.5 Hz > or < 59.3 Hz
 - R-L-C Load (IEEE 929 & UL 1741)
 - Quality factor $Q_f = 2.5 \rightarrow Q_L \ \& \ Q_C = 2.5 \times P_{\text{inv}}$
 - Calculated Range of K_a : **$0.076 < K_a < 0.3$**
 - $\eta_{\text{preset}} = 0.1$

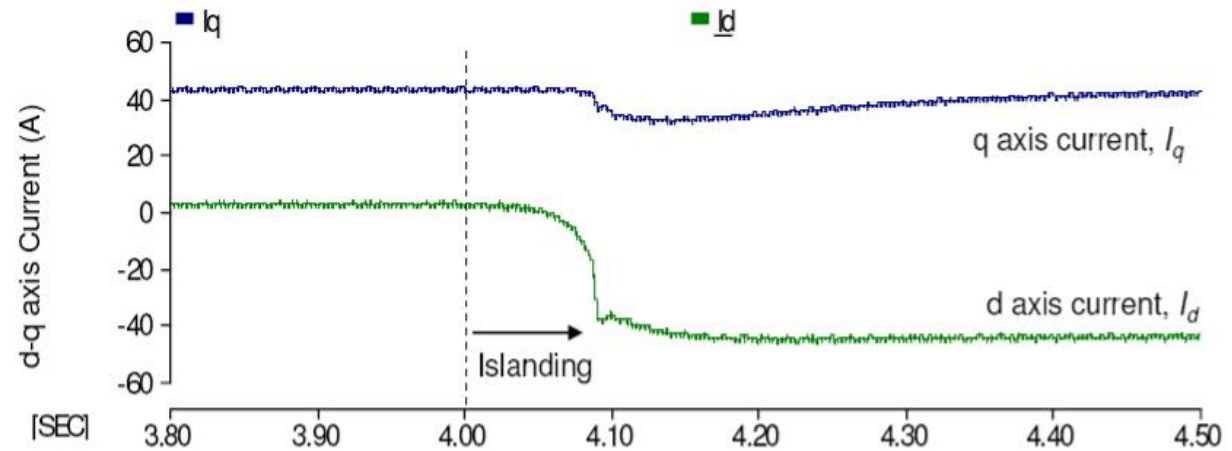
Without FSAC



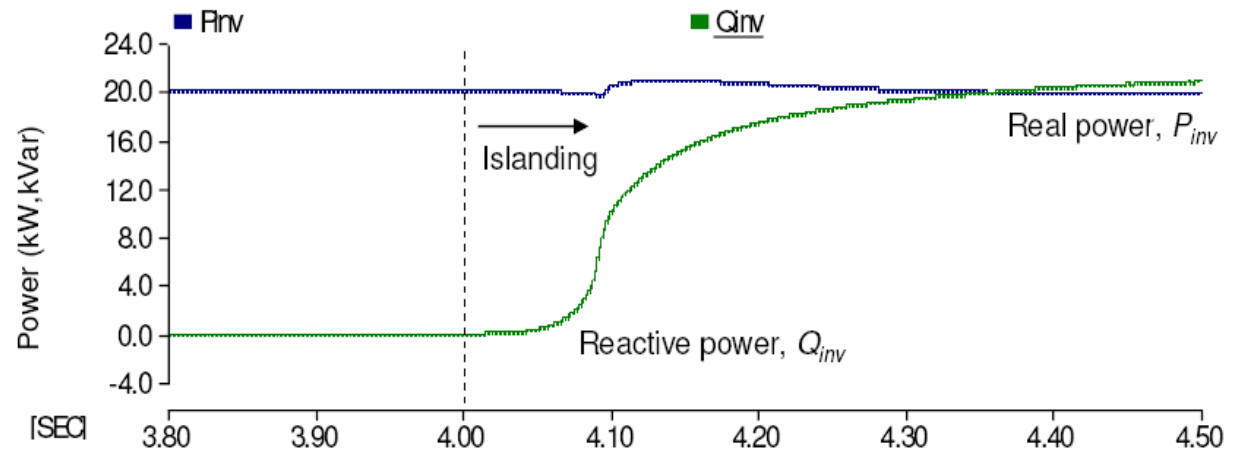
With FSAC ($K_a = 0.15$)



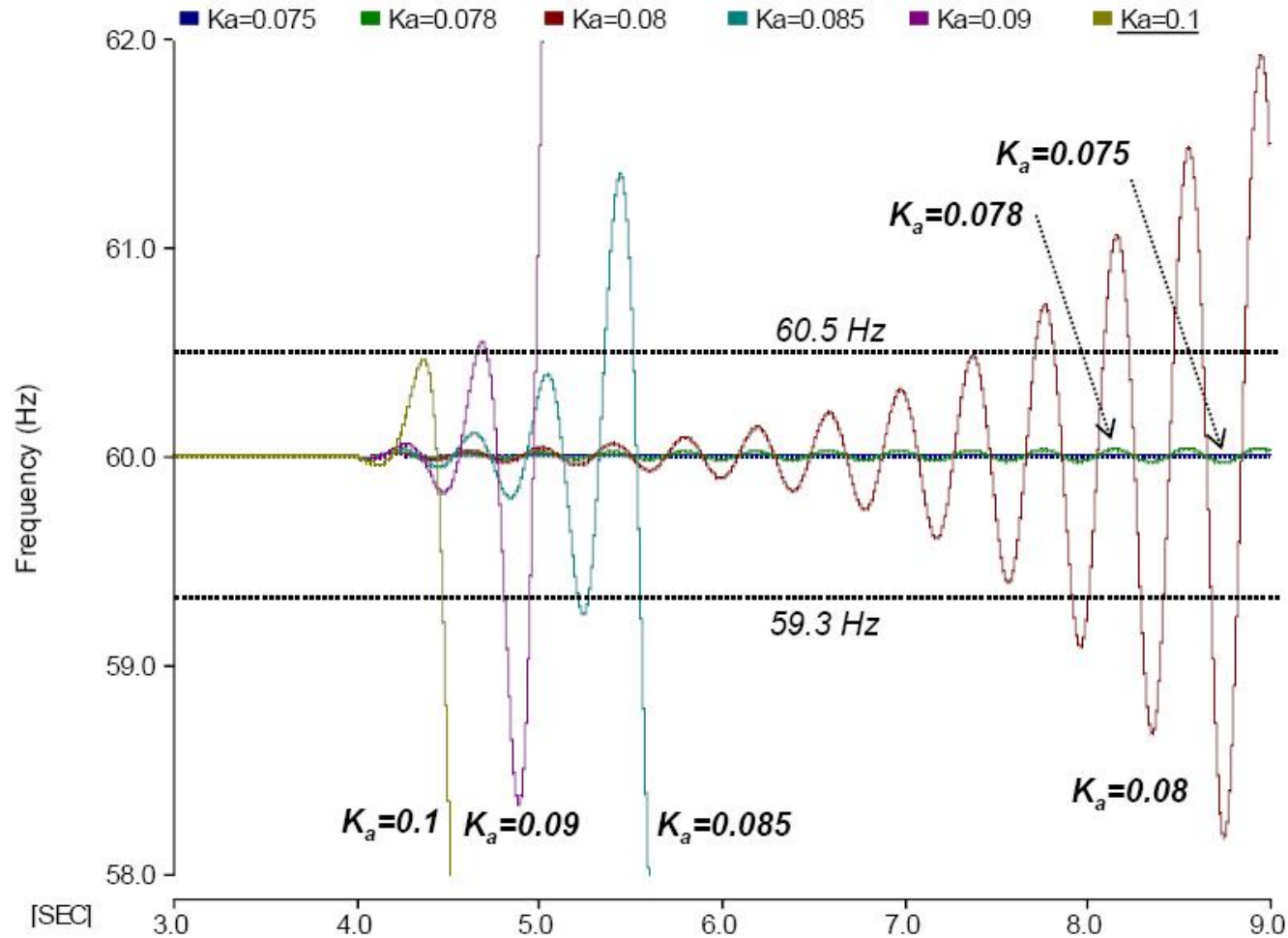
q-and d-axis current



Real and Reactive Power



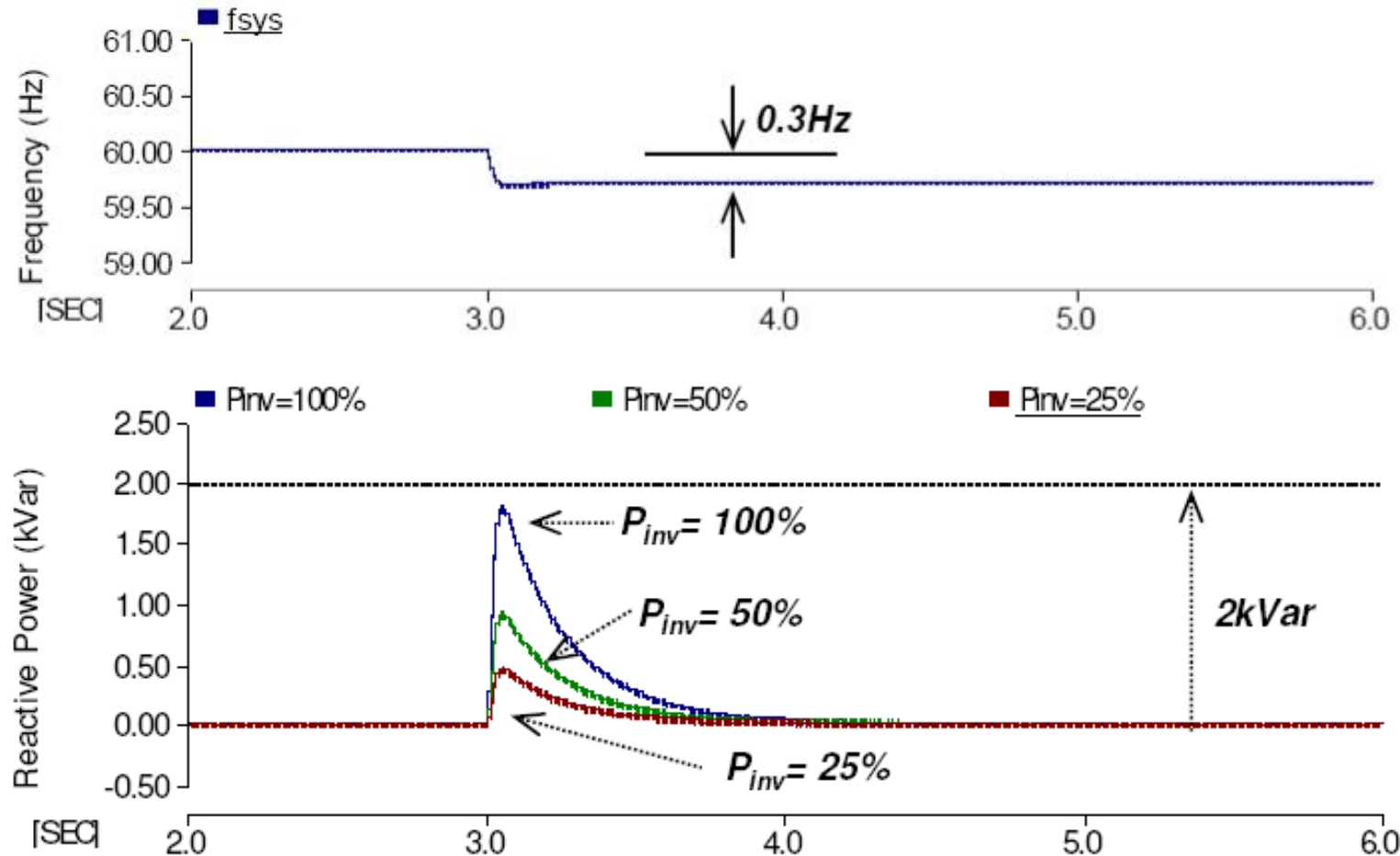
Frequency Variations with Different gains of K_a



Calculated & Simulated results for Lower Limit of K_a

| K_p | K_i | Calculated x100 | Simulatedx100 |
|-------|-------|-----------------|---------------|
| 2 | 5 | 2.6 | 2.8 |
| | 20 | 2.6 | 3.0 |
| | 50 | 2.6 | 3.3 |
| 5 | 5 | 4.5 | 4.6 |
| | 20 | 4.5 | 4.7 |
| | 50 | 4.5 | 5.0 |
| 10 | 5 | 7.6 | 7.6 |
| | 20 | 7.6 | 7.7 |
| | 50 | 7.6 | 7.8 |
| 15 | 5 | 10.6 | 10.6 |
| | 20 | 10.7 | 10.7 |
| | 50 | 10.7 | 10.7 |
| 20 | 5 | 13.7 | 13.6 |
| | 20 | 13.8 | 13.7 |
| | 50 | 13.8 | 13.7 |

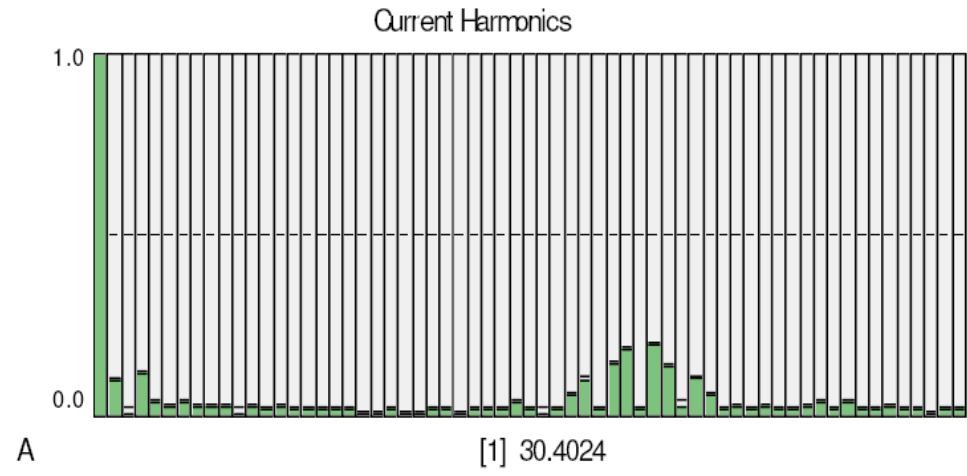
Q_{inv} disturbance due to $\Delta\omega$ in grid-tied operation



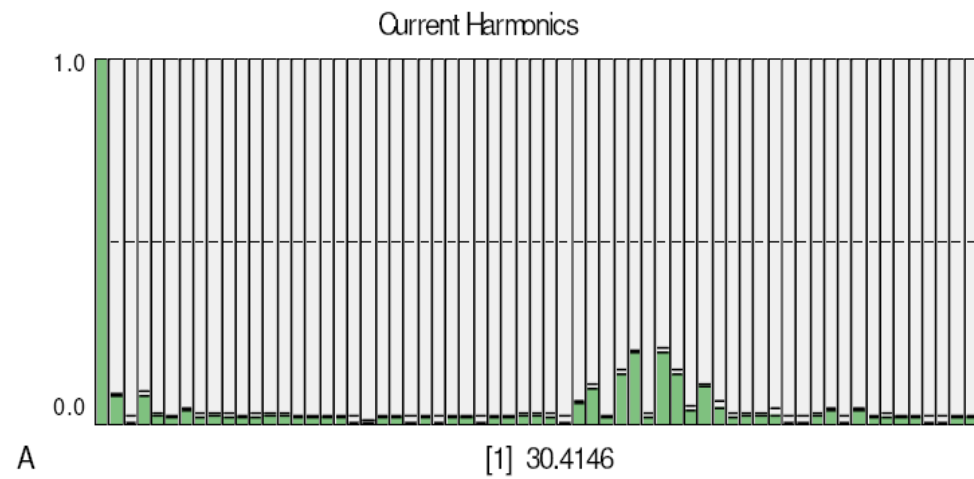
Q_{inv} at $K_a = 0.3$

Harmonic Spectrum

Without FSAC

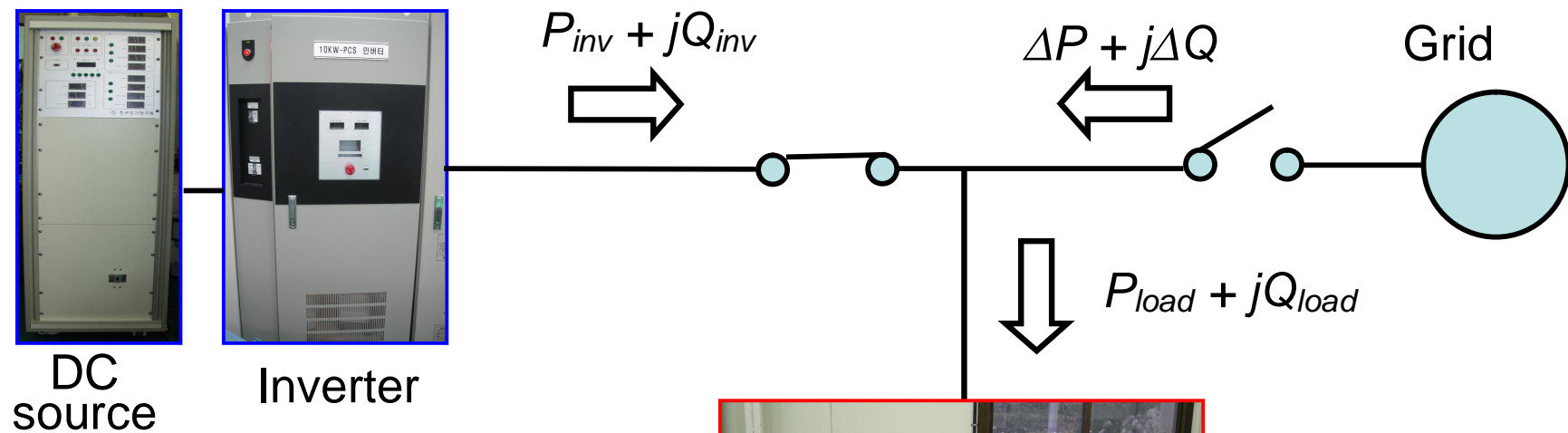


With FSAC



1st freq. → 63rd

Experimental Results



$$P_{inv} = 4.0\text{kW}, Q_{inv} = 0\text{kVar}$$

$$K_p = 10, K_i = 5$$

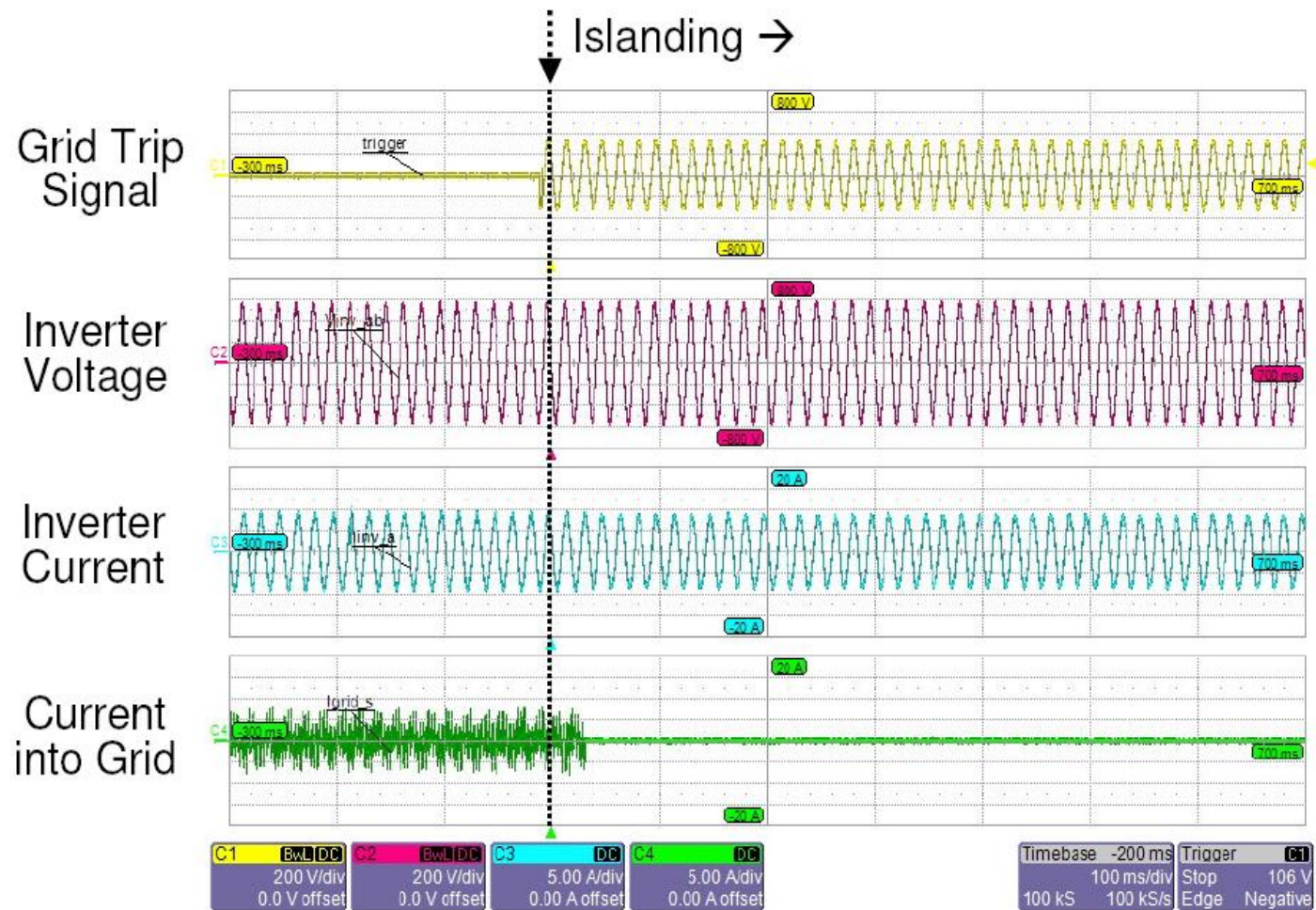
$$P_{load} = 4.0\text{kW}, Q_{load} = 0\text{kVar}$$

$$Q_f \text{ of RLC load} = 2.5$$

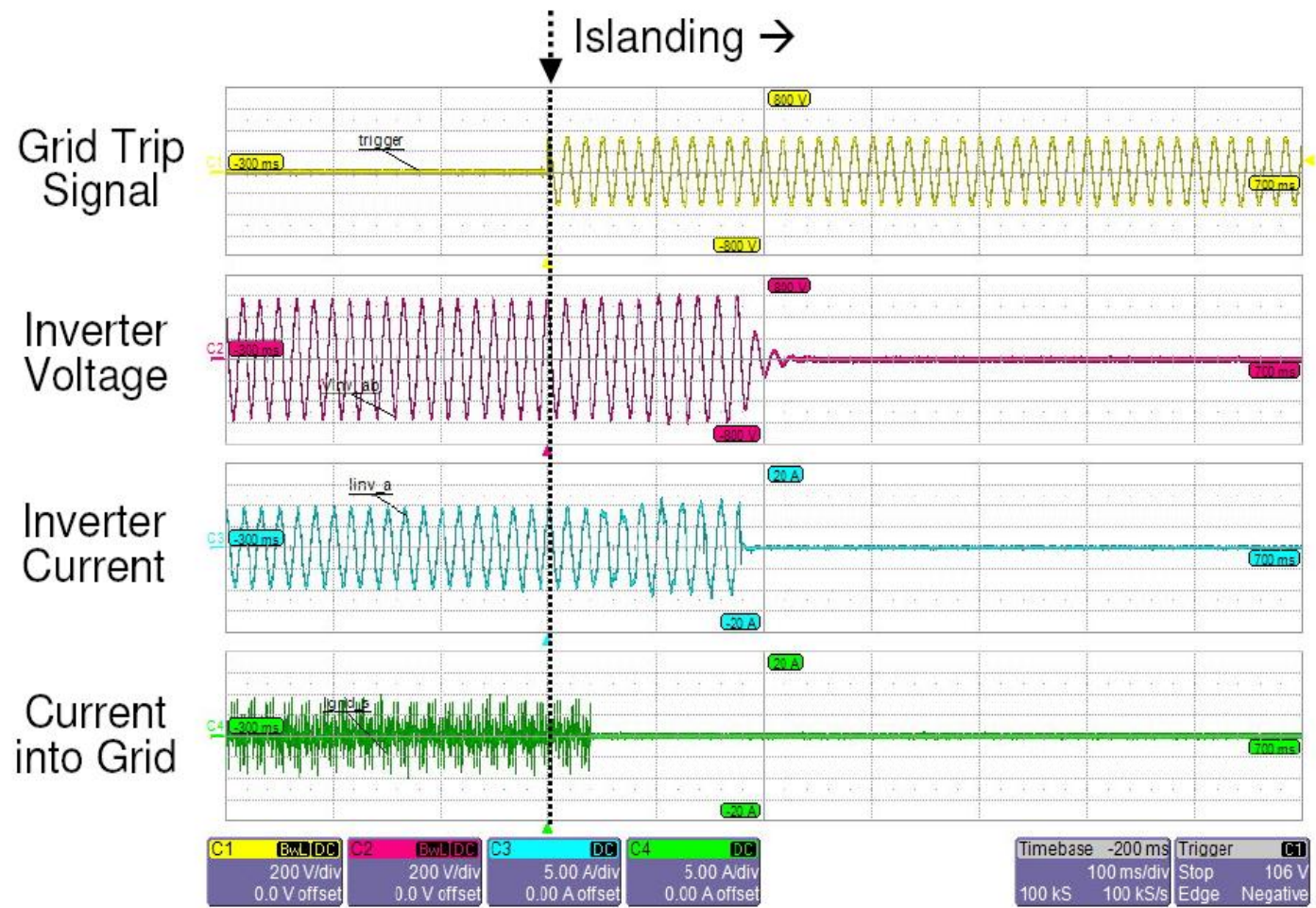


RLC Load

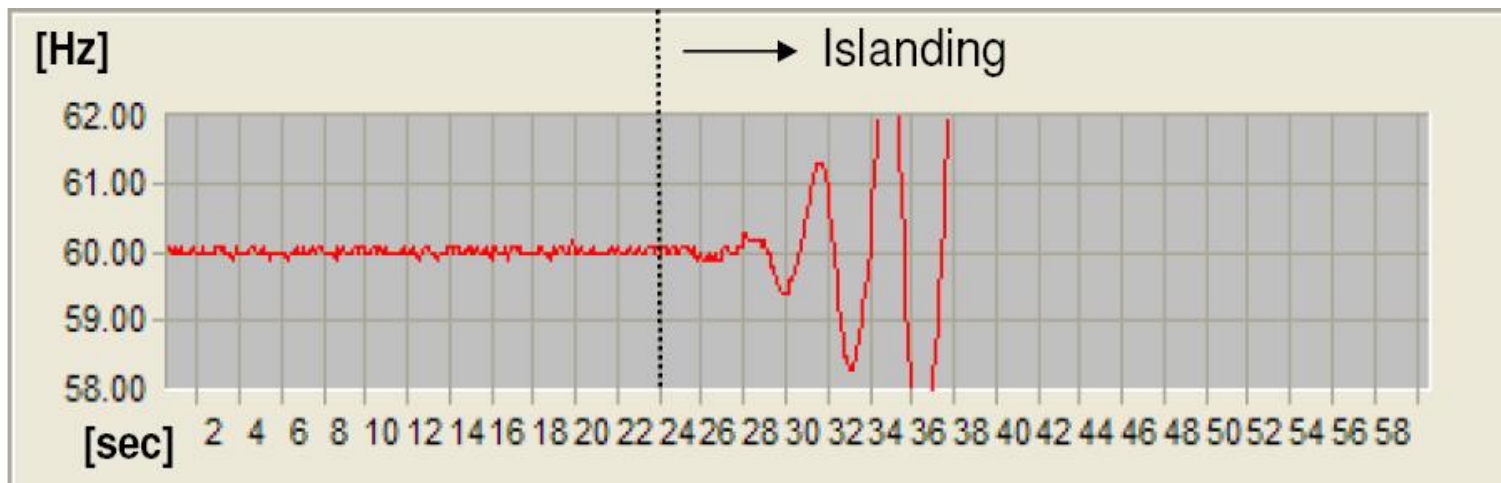
Before FSAC Implementation



After FSAC Implementation ($K_a = 0.1$)



Frequency with FSAC ($K_a = 0.057$)



Lower limit of K_a :

Calculation/simulation/experiment = 0.076/0.078/0.057

→ Acceptable

Conclusion

- Based on dq control and positive feedback
- P_{inv} dependency of control gain removed
- Design method and criteria suggested
- FSAC enables
 - Zero NDZ possible
 - Minimizing impact on power quality
 - Easy implementation